

(5400 + 5400) PIXELS \times 3 COLOR CCD LINEAR IMAGE SENSOR**DESCRIPTION**

The μ PD8872 is a color CCD (Charge Coupled Device) linear image sensor which changes optical images to electrical signal and has the function of color separation.

The μ PD8872 has 3 rows of (5400 + 5400) staggered pixels, and each row has a dual-sided readout-type charge transfer register. And it has reset feed-through level clamp circuits and voltage amplifiers. Therefore, it is suitable for 1200 dpi/A4 color image scanners, color facsimiles and so on.

FEATURES

- Valid photocell : (5400 + 5400) staggered pixels \times 3
- Photocell pitch : 5.25 μ m
- Line spacing : 63 μ m (12 lines) Red line - Green line, Green line - Blue line
10.5 μ m (2 lines) Odd line - Even line (for each color)
- Color filter : Primary colors (red, green and blue), pigment filter (with light resistance 10^7 lx \cdot hour)
- Resolution : 48 dot/mm A4 (210 \times 297 mm) size (shorter side)
1200 dpi US letter (8.5" \times 11") size (shorter side)
- Drive clock level : CMOS output under 5 V operation
- Data rate : 10 MHz Max.
- Power supply : +12 V
- On-chip circuits : Reset feed-through level clamp circuits
Voltage amplifiers

ORDERING INFORMATION

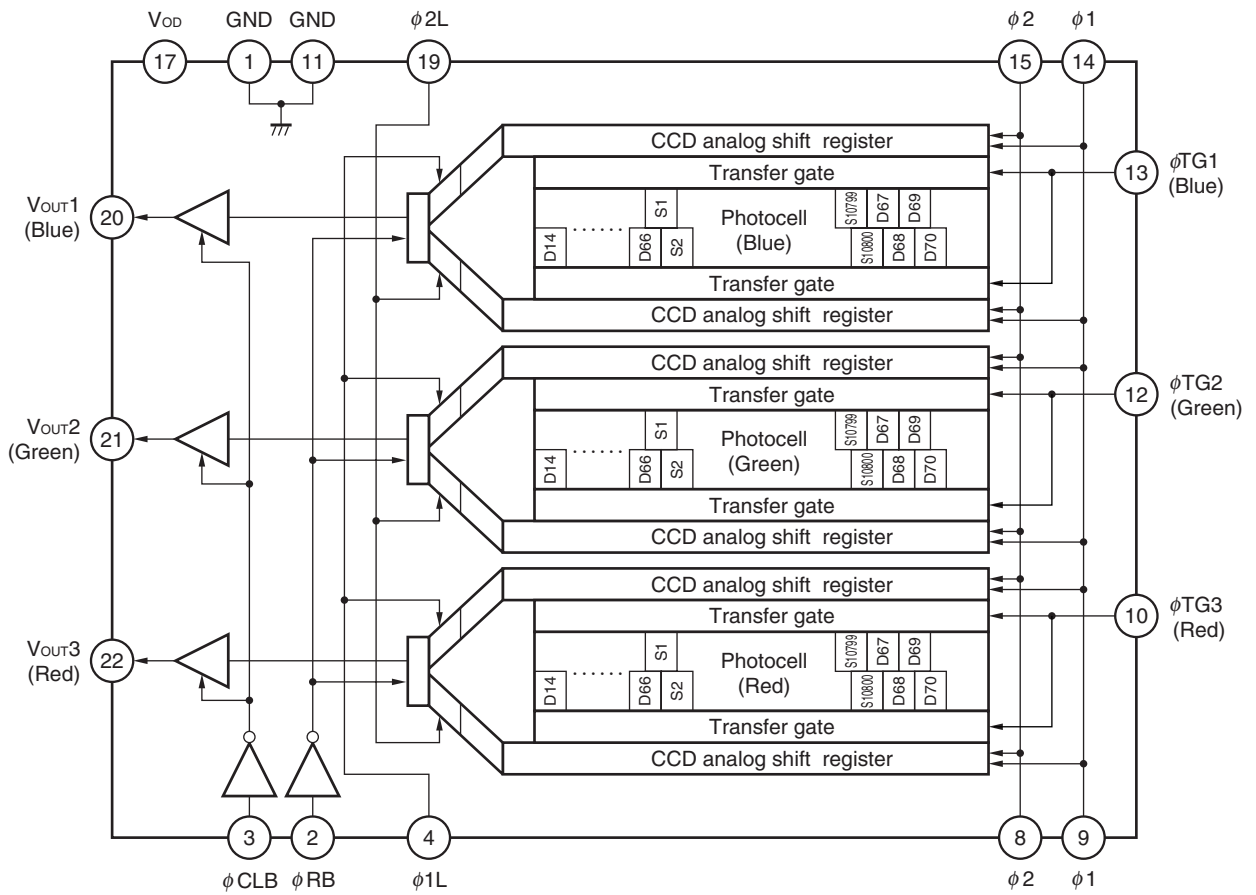
| Part Number | Package |
|------------------|---|
| μ PD8872CY-A | CCD linear image sensor 22-pin plastic DIP (10.16 mm (400)) |

<R> **Remark** The μ PD8872CY-A is a lead-free product.

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Not all products and/or types are available in every country. Please check with an NEC Electronics sales representative for availability and additional information.

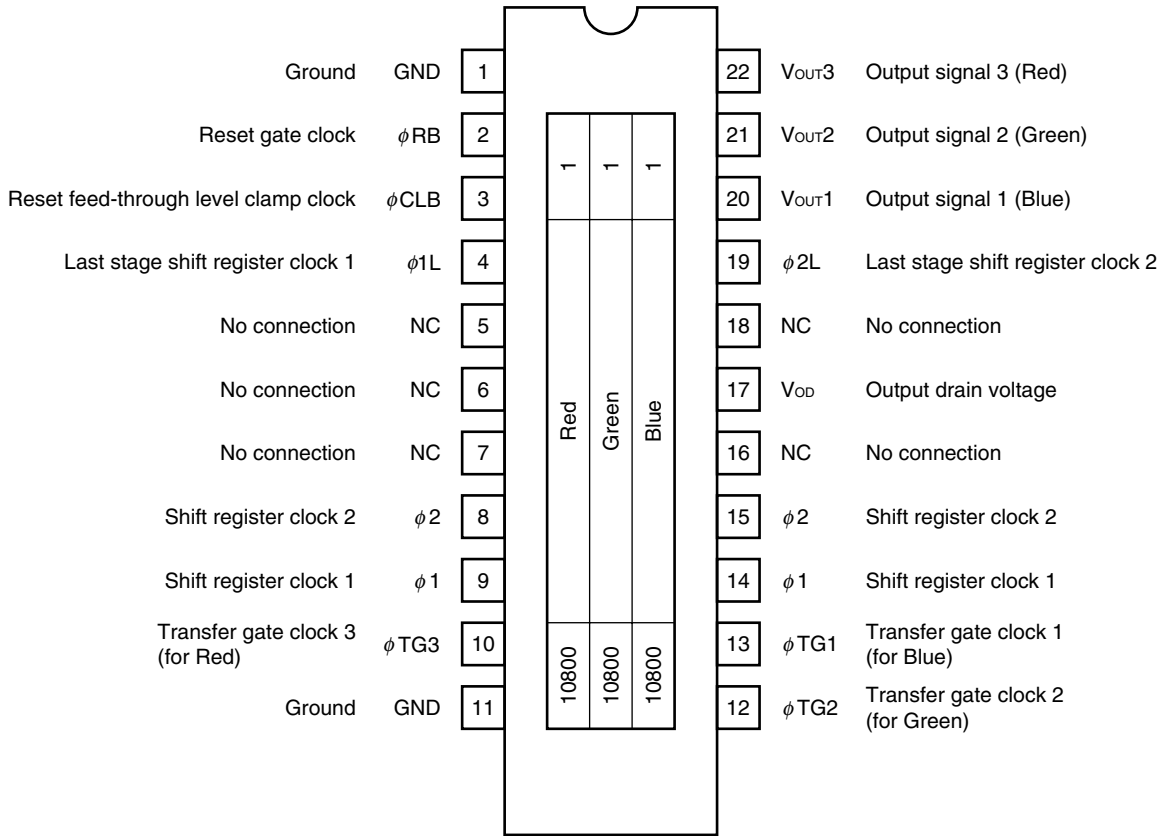
BLOCK DIAGRAM



PIN CONFIGURATION (Top View)

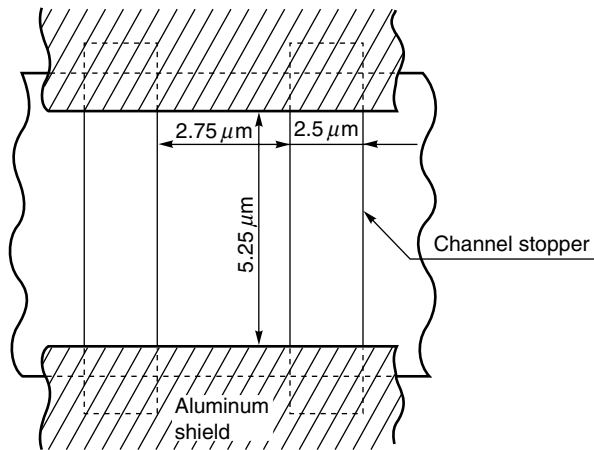
CCD linear image sensor 22-pin plastic DIP (10.16 mm (400))

• μPD8872CY-A

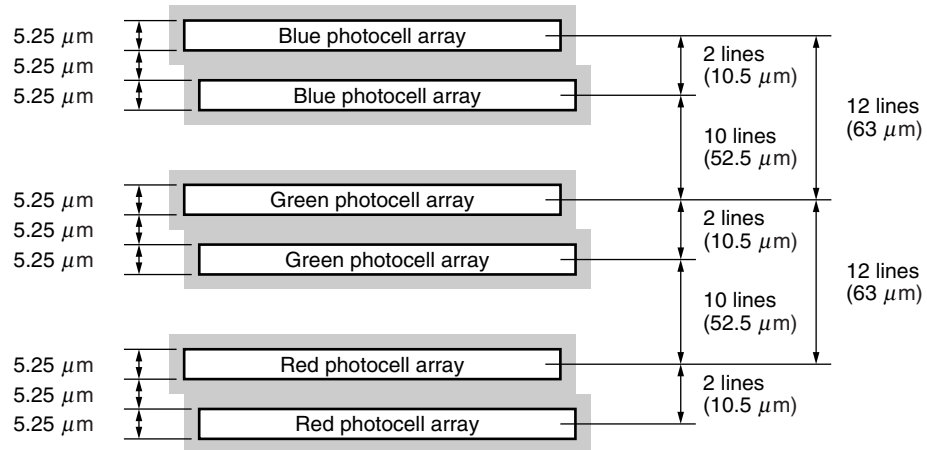


Caution Connect the No connection pins (NC) to GND.

PHOTOCELL STRUCTURE DIAGRAM



PHOTOCELL ARRAY STRUCTURE DIAGRAM (Line spacing)



ABSOLUTE MAXIMUM RATINGS (T_A = +25°C)

| Parameter | Symbol | Ratings | Unit |
|---|---|-------------|------|
| Output drain voltage | V _{OD} | -0.3 to +15 | V |
| Shift register clock voltage | V _{φ1} , V _{φ2} , V _{φ1L} , V _{φ2L} | -0.3 to +8 | V |
| Reset gate clock voltage | V _{φRB} | -0.3 to +8 | V |
| Reset feed-through level clamp clock voltage | V _{φCLB} | -0.3 to +8 | V |
| Transfer gate clock voltage | V _{φTG1} to V _{φTG3} | -0.3 to +8 | V |
| Operating ambient temperature ^{Note} | T _A | 0 to +60 | °C |
| Storage temperature | T _{stg} | -40 to +70 | °C |

Note Use at the condition without dew condensation.

Caution Product quality may suffer if the absolute maximum rating is exceeded even momentarily for any parameter. That is, the absolute maximum ratings are rated values at which the product is on the verge of suffering physical damage, and therefore the product must be used under conditions that ensure that the absolute maximum ratings are not exceeded.

RECOMMENDED OPERATING CONDITIONS (T_A = +25°C)

| Parameter | Symbol | Min. | Typ. | Max. | Unit |
|---|---|------|-----------------------------------|-----------------------------------|------|
| Output drain voltage | V _{OD} | 11.4 | 12.0 | 12.6 | V |
| Shift register clock high level | V _{φ1_H} , V _{φ2_H} , V _{φ1LH} , V _{φ2LH} | 4.75 | 5.0 | 5.5 | V |
| Shift register clock low level | V _{φ1_L} , V _{φ2_L} , V _{φ1LL} , V _{φ2LL} | -0.3 | 0 | +0.25 | V |
| Reset gate clock high level | V _{φRBH} | 4.5 | 5.0 | 5.5 | V |
| Reset gate clock low level | V _{φRBL} | -0.3 | 0 | +0.5 | V |
| Reset feed-through level clamp clock high level | V _{φCLBH} | 4.5 | 5.0 | 5.5 | V |
| Reset feed-through level clamp clock low level | V _{φCLBL} | -0.3 | 0 | +0.5 | V |
| Transfer gate clock high level | V _{φTG1H} to V _{φTG3H} | 4.75 | V _{φ1_H} ^{Note} | V _{φ1_H} ^{Note} | V |
| Transfer gate clock low level | V _{φTG1L} to V _{φTG3L} | -0.3 | 0 | +0.15 | V |
| Data rate | f _{φRB} | - | 2.0 | 10.0 | MHz |

Note When Transfer gate clock high level (V_{φTG1H} to V_{φTG3H}) is higher than Shift register clock high level (V_{φ1_H}), Image lag can increase.

ELECTRICAL CHARACTERISTICS

$T_A = +25^\circ\text{C}$, $V_{OD} = 12\text{ V}$, data rate ($f_{\phi RB}$) = 2 MHz, storage time = 5.5 ms, input signal clock = 5 V_{p-p},
light source : 3200 K halogen lamp + C-500S (infrared cut filter, t = 1 mm) + HA-50 (heat absorbing filter, t = 3 mm)

| Parameter | Symbol | Test Conditions | Min. | Typ. | Max. | Unit |
|--|------------------|--|-------|-------|-------|--------|
| Saturation voltage | V _{sat} | | 2.7 | 3.0 | – | V |
| Saturation exposure | Red | SER | – | 0.505 | – | lx•s |
| | Green | SEG | – | 0.573 | – | lx•s |
| | Blue | SEB | – | 0.888 | – | lx•s |
| Photo response non-uniformity | PRNU | V _{OUT} = 1.0 V | – | 6 | 20 | % |
| Average dark signal | ADS | Light shielding | – | 0.2 | 2.0 | mV |
| Dark signal non-uniformity | DSNU | Light shielding | – | 1.5 | 5.0 | mV |
| Power consumption | P _w | | – | 360 | 540 | mW |
| Output impedance | Z _o | | – | 0.35 | 1.00 | kΩ |
| Response | Red | R _R | 4.15 | 5.94 | 7.73 | V/lx•s |
| | Green | R _G | 3.66 | 5.24 | 6.82 | V/lx•s |
| | Blue | R _B | 2.36 | 3.38 | 4.39 | V/lx•s |
| Image lag | IL | V _{OUT} = 1.0 V | – | 3.0 | 7.0 | % |
| Offset level ^{Note 1} | V _{OS} | | 4.5 | 6.0 | 7.5 | V |
| Output fall delay time ^{Note 2} | t _d | V _{OUT} = 1.0 V, t1', t2' = 5 ns | – | 25 | – | ns |
| Total transfer efficiency | TTE | V _{OUT} = 1.0 V, data rate = 10 MHz | 92 | 98 | – | % |
| Register imbalance | RI | V _{OUT} = 1.0 V | – | 1.0 | 4.0 | % |
| Response peak | Red | | – | 630 | – | nm |
| | Green | | – | 540 | – | nm |
| | Blue | | – | 460 | – | nm |
| Dynamic range | DR1 | V _{sat} /DSNU | – | 2000 | – | times |
| | DR2 | V _{sat} /σCDS | – | 3000 | – | times |
| Reset feed-through noise ^{Notes 1, 3} | RFTN | Light shielding | –2000 | +300 | +1000 | mV |
| Random noise (CDS) | σCDS | Light shielding | – | 1.0 | – | mV |

Notes 1. Refer to **TIMING CHART 2, 3.**

2. When the fall time of φ1L and φ2L (t1', t2') is the Typ. value (refer to **TIMING CHART 2, 3**).

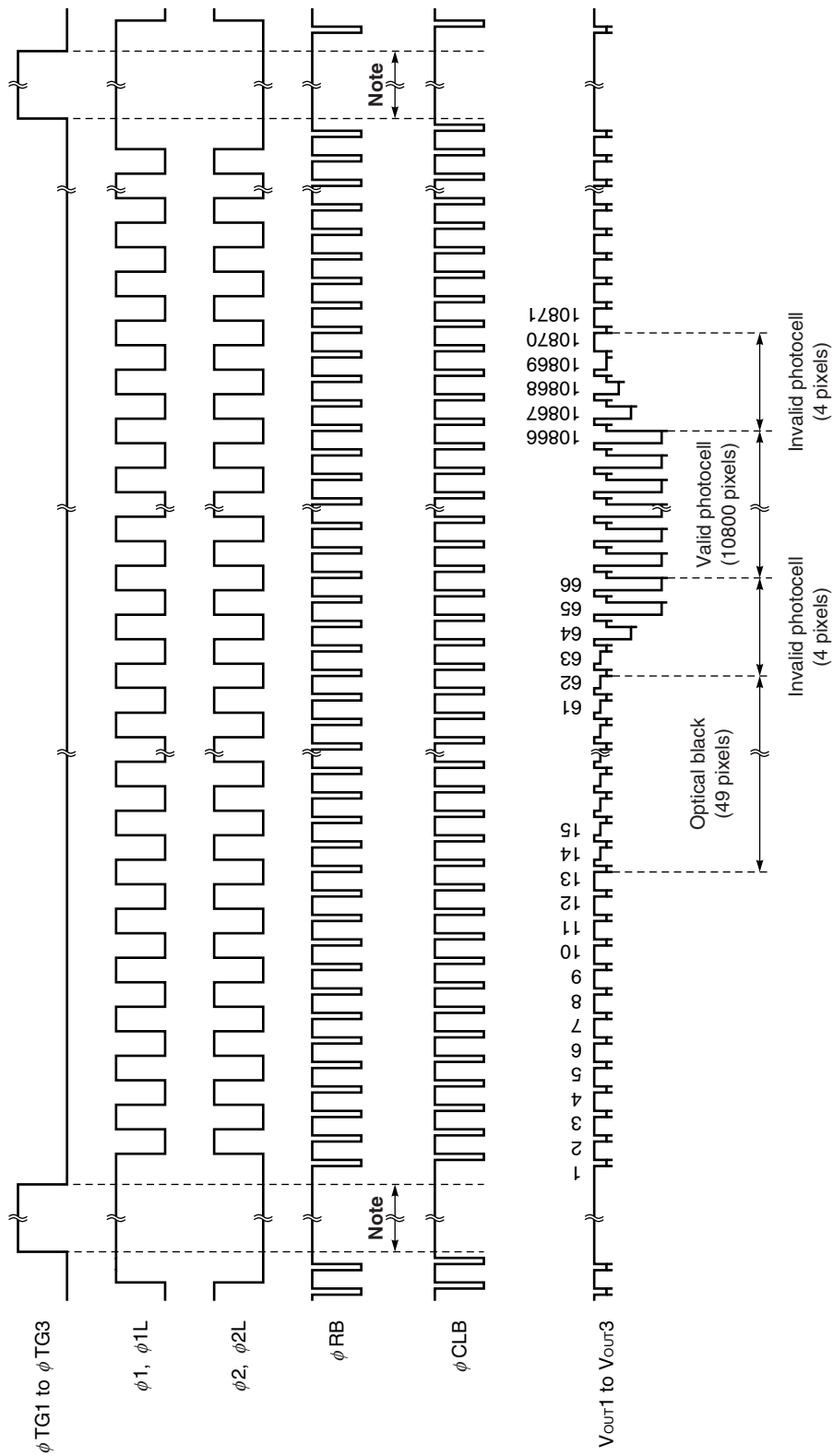
3. The reset feed-through noise changes by peripheral circuit of register, the driver circuit for φRB and so on.

INPUT PIN CAPACITANCE (T_A = +25°C, V_{OD} = 12 V)

| Parameter | Symbol | Pin name | Pin No. | Min. | Typ. | Max. | Unit |
|--|-----------------------|----------|---------|------|------|------|------|
| Shift register clock pin capacitance 1 | C _{φ1} | φ 1 | 9 | – | 500 | – | pF |
| | | | 14 | – | 500 | – | pF |
| | φ 1 total capacitance | | | – | 1000 | – | pF |
| Shift register clock pin capacitance 2 | C _{φ2} | φ 2 | 8 | – | 500 | – | pF |
| | | | 15 | – | 500 | – | pF |
| | φ 2 total capacitance | | | – | 1000 | – | pF |
| Last stage shift register clock pin capacitance | C _{φL} | φ 1L | 4 | – | 10 | – | pF |
| | | φ 2L | 19 | – | 10 | – | pF |
| Reset gate clock pin capacitance | C _{φRB} | φ RB | 2 | – | 10 | – | pF |
| Reset feed-through level clamp clock pin capacitance | C _{φCLB} | φ CLB | 3 | – | 10 | – | pF |
| Transfer gate clock pin capacitance | C _{φTG} | φ TG1 | 13 | – | 200 | – | pF |
| | | φ TG2 | 12 | – | 200 | – | pF |
| | | φ TG3 | 10 | – | 200 | – | pF |

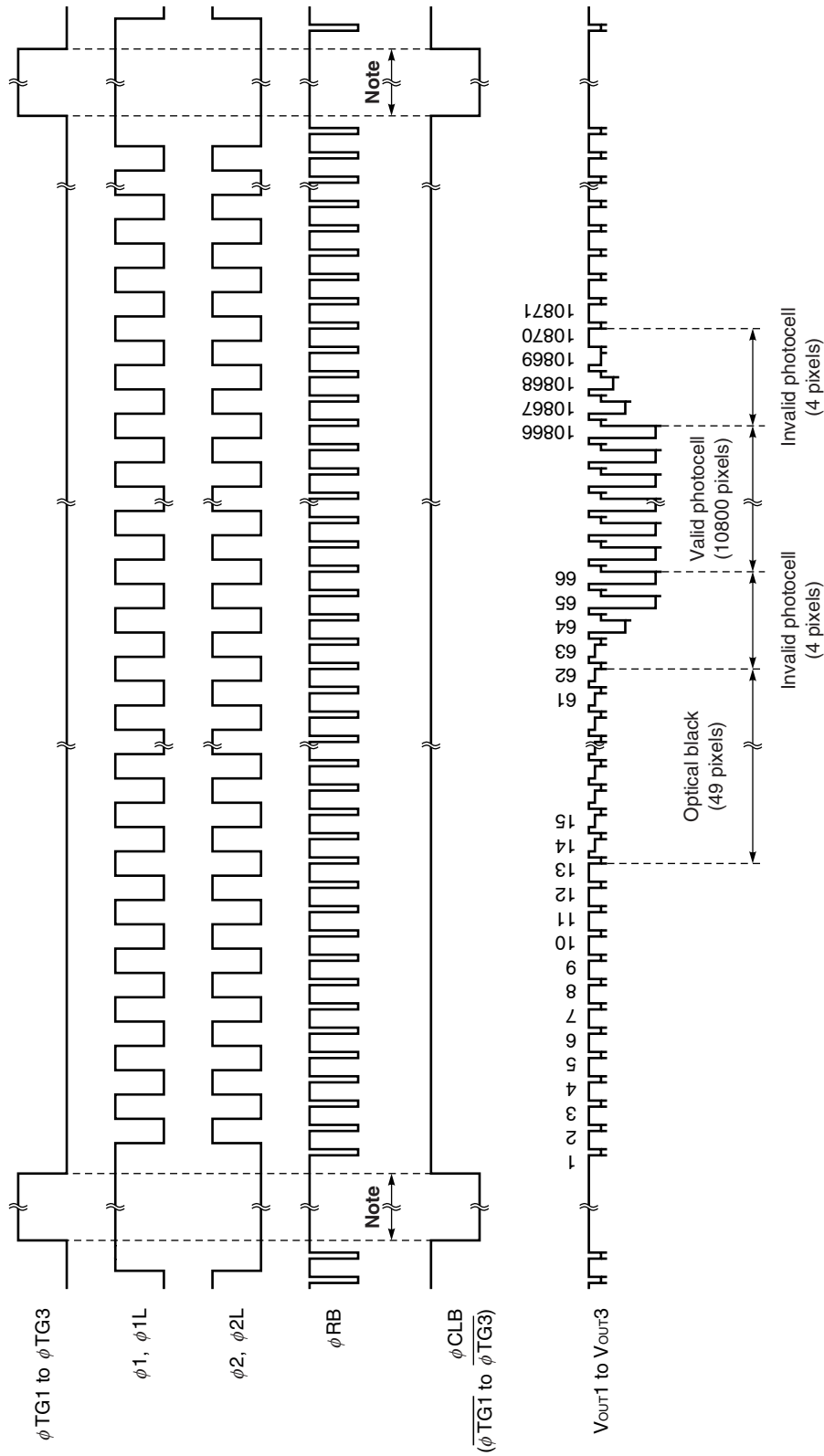
- Remarks 1.** Pins 9 and 14 (φ1), 8 and 15 (φ2) are each connected inside of the device.
- 2.** C_{φ1} and C_{φ2} show the equivalent capacity of the real drive including the capacity of between φ 1 and φ 2.

TIMING CHART 1-1 (Bit clamp mode, for each color)



Note Set the ϕ RB and ϕ CLB to high level during this period.

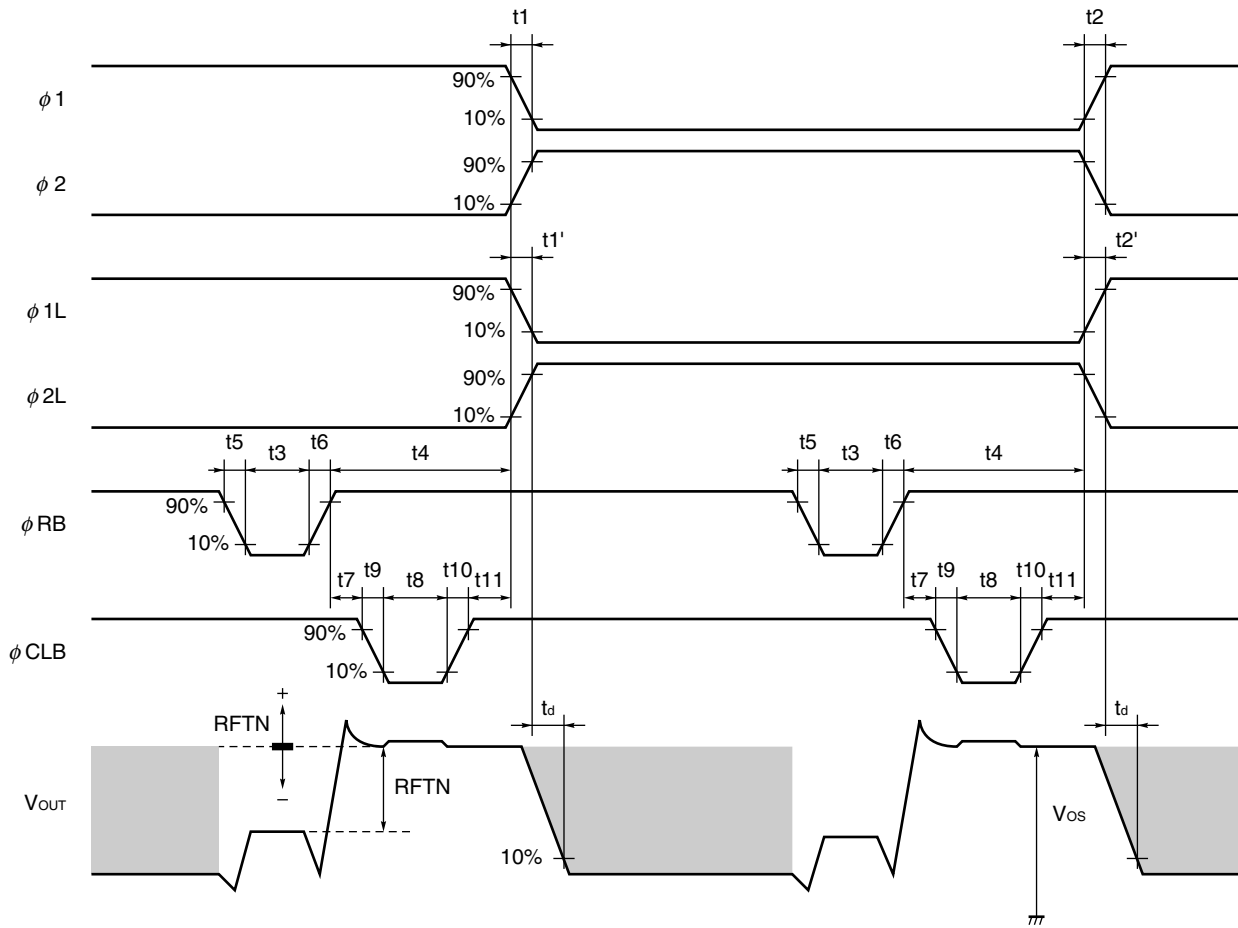
TIMING CHART 1-2 (Line clamp mode, for each color)



Note Set the φRB to high level during this period.

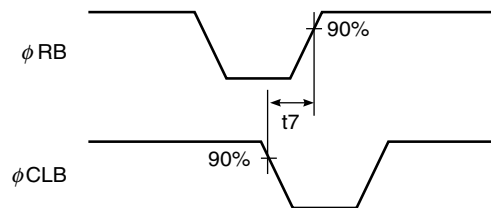
Remark Inverse pulse of the φTG1 to φTG3 can be used as φCLB.

TIMING CHART 2 (Bit clamp mode, for each color)

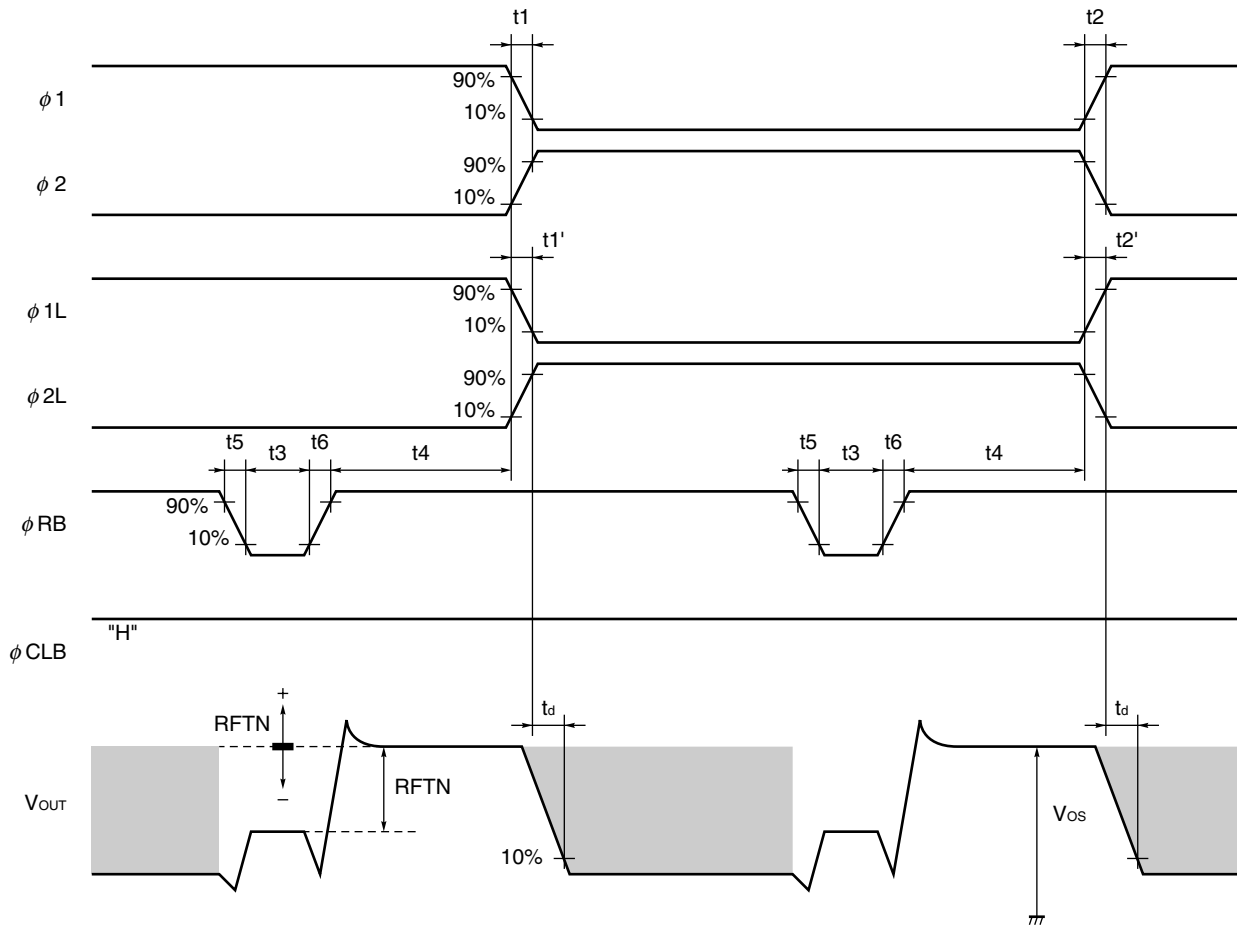


| Symbol | Min. | Typ. | Max. | Unit |
|----------|--------------------|------|------|------|
| t1, t2 | 0 | 25 | – | ns |
| t1', t2' | 0 | 5 | – | ns |
| t3 | 20 | 100 | – | ns |
| t4 | 30 | 150 | – | ns |
| t5, t6 | 0 | 25 | – | ns |
| t7 | -5 ^{Note} | 25 | – | ns |
| t8 | 20 | 100 | – | ns |
| t9, t10 | 0 | 25 | – | ns |
| t11 | 5 | 25 | – | ns |

Note Min. of t7 shows that the ϕRB and ϕCLB overlap each other.

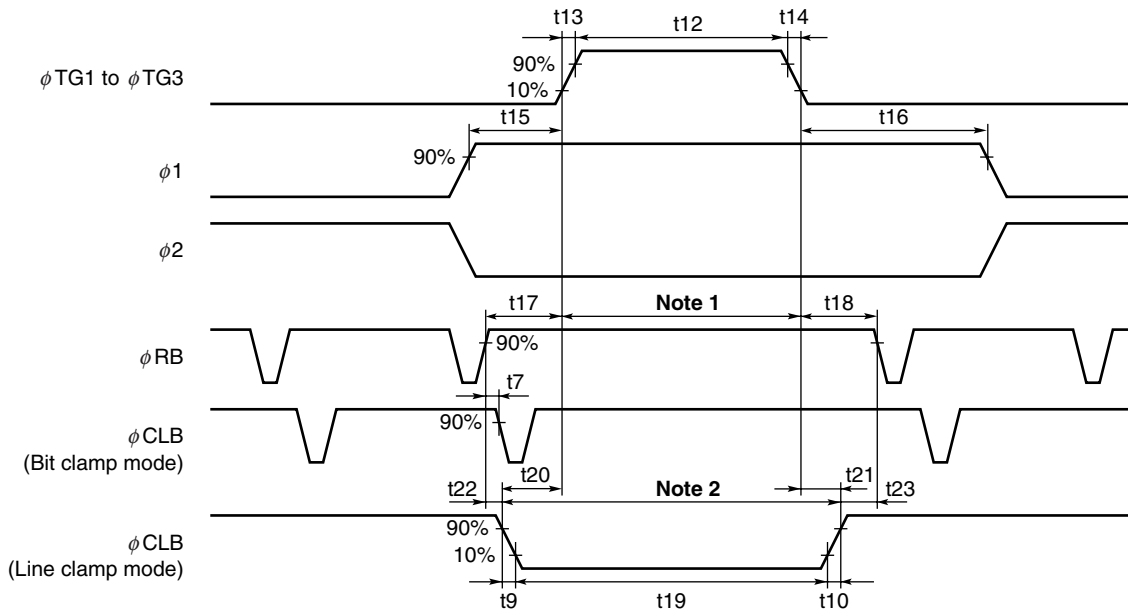


TIMING CHART 3 (Line clamp mode, for each color)



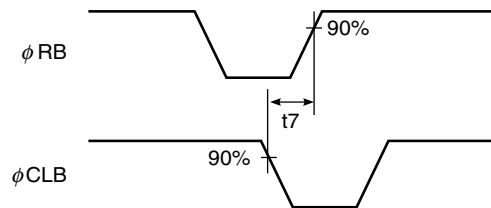
| Symbol | Min. | Typ. | Max. | Unit |
|----------|------|------|------|------|
| t1, t2 | 0 | 25 | – | ns |
| t1', t2' | 0 | 5 | – | ns |
| t3 | 20 | 100 | – | ns |
| t4 | 30 | 150 | – | ns |
| t5, t6 | 0 | 25 | – | ns |

φTG1 to φTG3, φ1, φ2 TIMING CHART



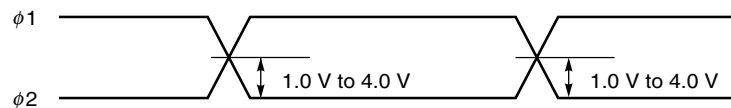
| Symbol | Min. | Typ. | Max. | Unit |
|----------|----------------------|-------|-------|------|
| t7 | -5 ^{Note 3} | 25 | - | ns |
| t9, t10 | 0 | 25 | - | ns |
| t12 | 5000 | 10000 | 50000 | ns |
| t13, t14 | 0 | 50 | - | ns |
| t15, t16 | 900 | 1000 | - | ns |
| t17, t18 | 200 | 400 | - | ns |
| t19 | t12 | t12 | 50000 | ns |
| t20, t21 | 0 | 50 | - | ns |
| t22, t23 | 0 | 350 | - | ns |

- Notes 1.** Set the φRB and φCLB to high level during this period.
- 2.** Set the φRB to high level during this period.
- 3.** Min. of t7 shows that the φRB and φCLB overlap each other.

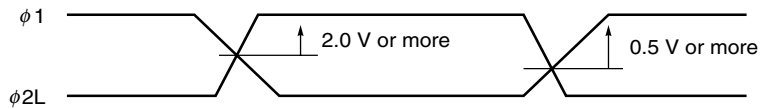


Remark Inverse pulse of the φTG1 to φTG3 can be used as φCLB.

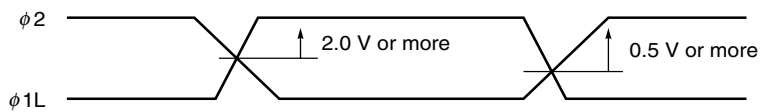
$\phi 1, \phi 2$ cross points



$\phi 1, \phi 2L$ cross points



$\phi 2, \phi 1L$ cross points



Remark Adjust cross points ($\phi 1, \phi 2$), ($\phi 1, \phi 2L$) and ($\phi 2, \phi 1L$) with input resistance of each pin.

DEFINITIONS OF CHARACTERISTIC ITEMS

1. Saturation voltage : **V_{sat}**
Output signal voltage at which the response linearity is lost.

2. Saturation exposure : **SE**
Product of intensity of illumination (lx) and storage time (s) when saturation of output voltage occurs.

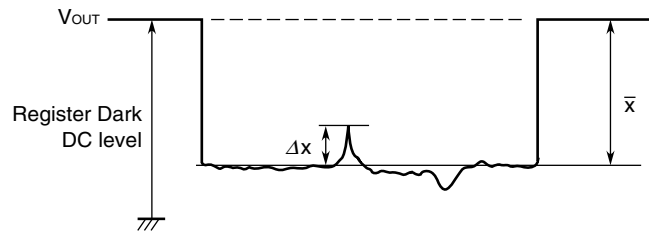
3. Photo response non-uniformity : **PRNU**
The output signal non-uniformity of all the valid pixels when the photosensitive surface is applied with the light of uniform illumination. This is calculated by the following formula.

$$PRNU (\%) = \frac{\Delta x}{\bar{x}} \times 100$$

Δx : maximum of |x_j - x̄|

$$\bar{x} = \frac{\sum_{j=1}^{10800} x_j}{10800}$$

x_j : Output voltage of valid pixel number j



4. Average dark signal : **ADS**
Average output signal voltage of all the valid pixels at light shielding. This is calculated by the following formula.

$$ADS (mV) = \frac{\sum_{j=1}^{10800} d_j}{10800}$$

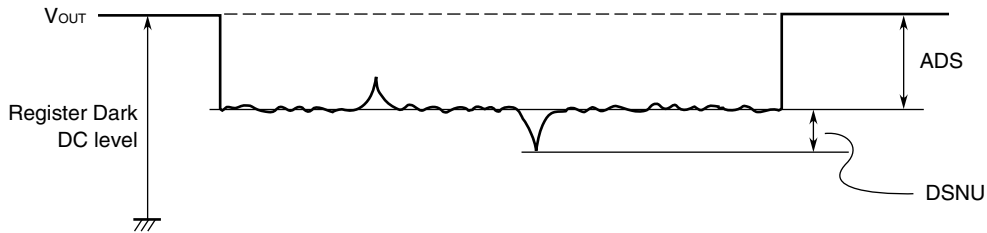
d_j : Dark signal of valid pixel number j

5. Dark signal non-uniformity : **DSNU**

Absolute maximum of the difference between ADS and voltage of the highest or lowest output pixel of all the valid pixels at light shielding. This is calculated by the following formula.

$$\text{DSNU (mV)} : \text{maximum of } |d_j - \text{ADS}|_{j=1 \text{ to } 10800}$$

d_j : Dark signal of valid pixel number j



6. Output impedance : **Zo**

Impedance of the output pins viewed from outside.

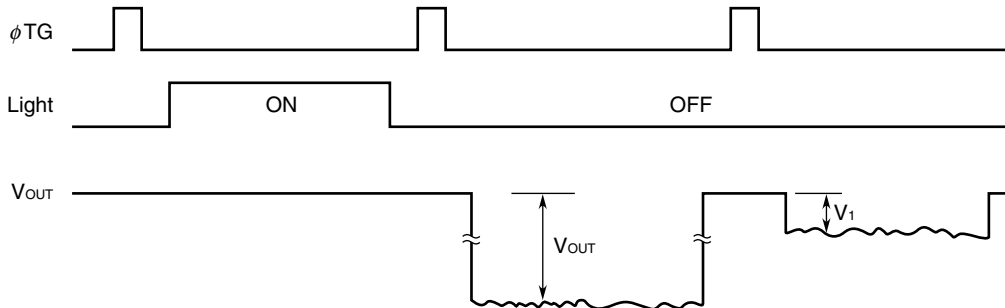
7. Response : **R**

Output voltage divided by exposure ($I \times s$).

Note that the response varies with a light source (spectral characteristic).

8. Image lag : **IL**

The rate between the last output voltage and the next one after read out the data of a line.



$$\text{IL (\%)} = \frac{V_1}{V_{\text{OUT}}} \times 100$$

9. Register imbalance: **RI**

The rate of the difference between the averages of the output voltage of Odd and Even pixels, against the average output voltage of all the valid pixels.

$$RI (\%) = \frac{\frac{2}{n} \left| \sum_{j=1}^{\frac{n}{2}} (V_{2j-1} - V_{2j}) \right|}{\frac{1}{n} \sum_{j=1}^n V_j} \times 100$$

n : Number of valid pixels
 V_j : Output voltage of each pixel

10. Random noise (CDS) : **σCDS**

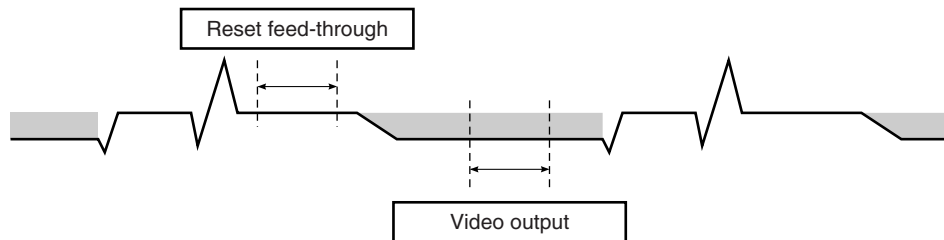
Random noise σCDS is defined as the standard deviation of a valid pixel output signal with 100 times (=100 lines) data sampling at dark (light shielding). σCDS is calculated by the following procedure.

1. One valid photocell in one reading is fixed as measurement point.
2. The output level is measured during the reset feed-through period which is averaged over 100 ns to get “VD_i”.
3. The output level is measured during the video output time averaged over 100 ns to get “VO_i”.
4. The correlated double sampling output is defined by the following formula.

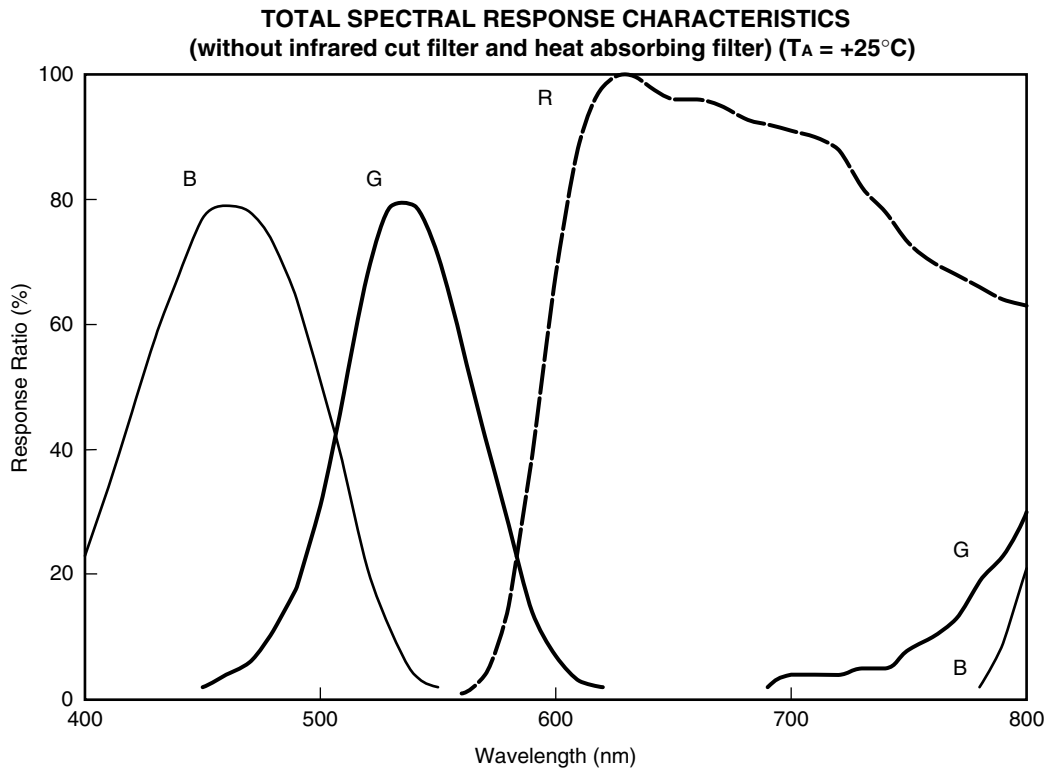
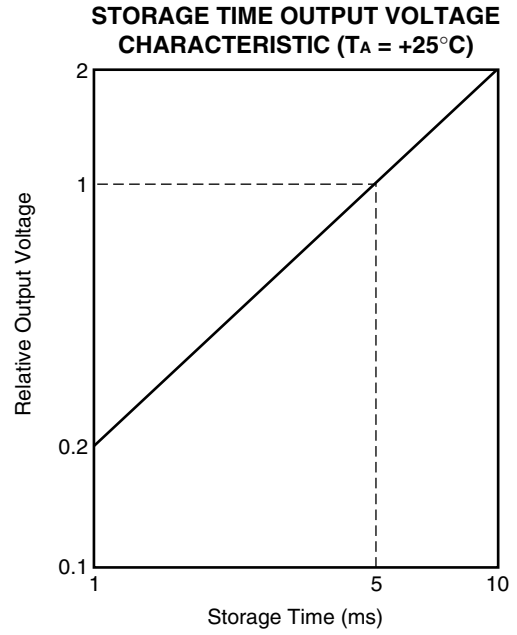
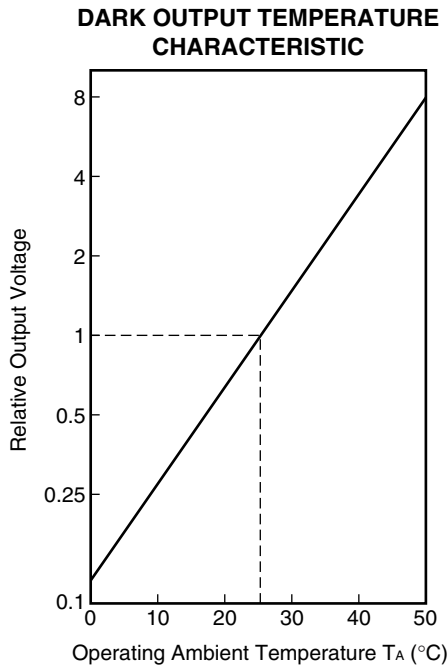
$$VCDS_i = VD_i - VO_i$$

5. Repeat the above procedure (1 to 4) for 100 times (= 100 lines).
6. Calculate the standard deviation σCDS using the following formula equation.

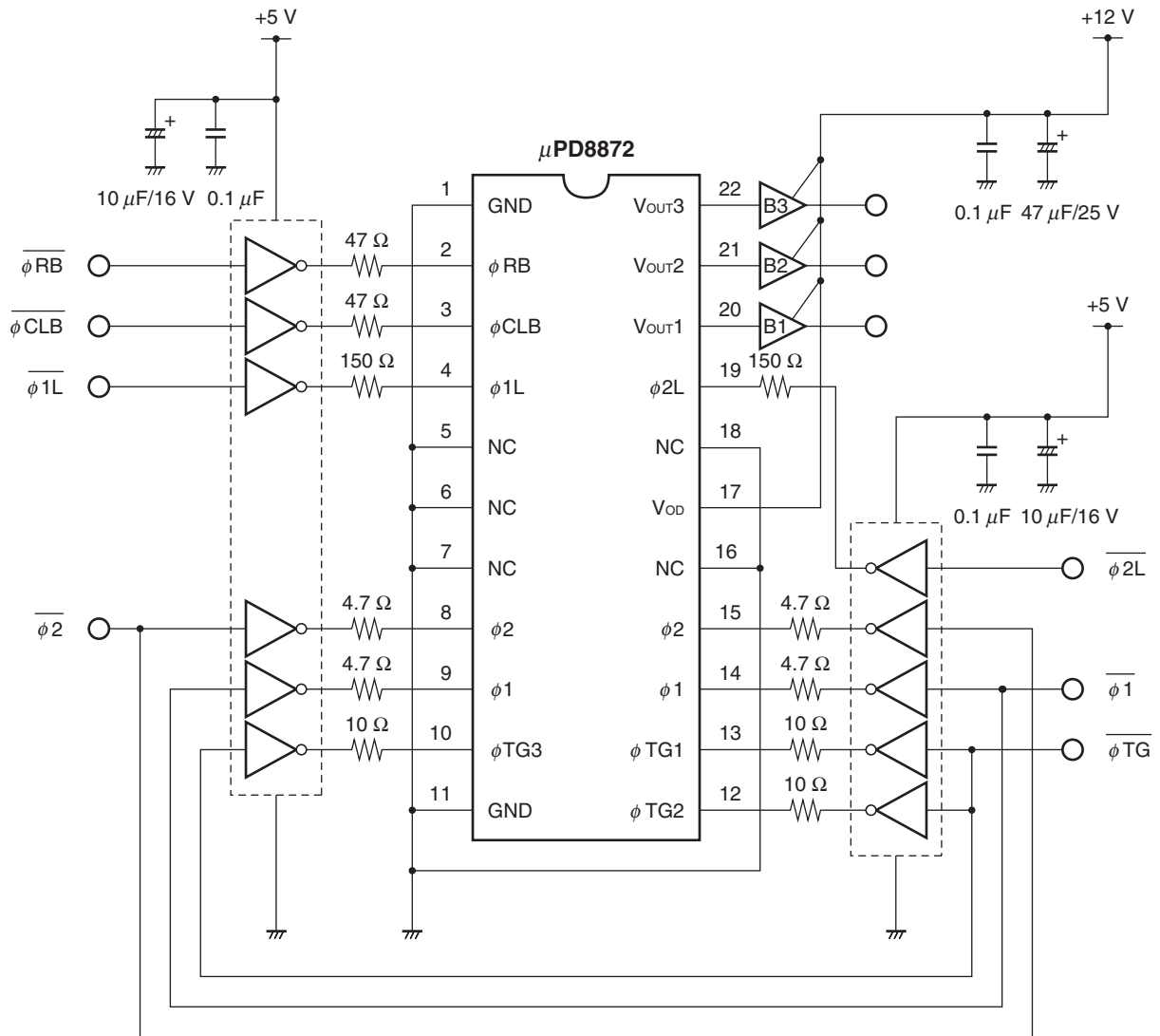
$$\sigma_{CDS} (mV) = \sqrt{\frac{\sum_{i=1}^{100} (VCDS_i - \bar{V})^2}{100}}, \quad \bar{V} = \frac{1}{100} \sum_{i=1}^{100} VCDS_i$$



STANDARD CHARACTERISTIC CURVES (Reference Value)

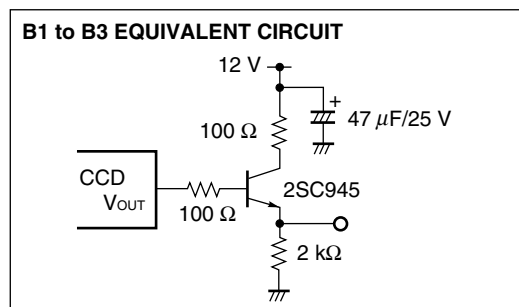


APPLICATION CIRCUIT EXAMPLE



Caution Connect the No connection pins (NC) to GND.

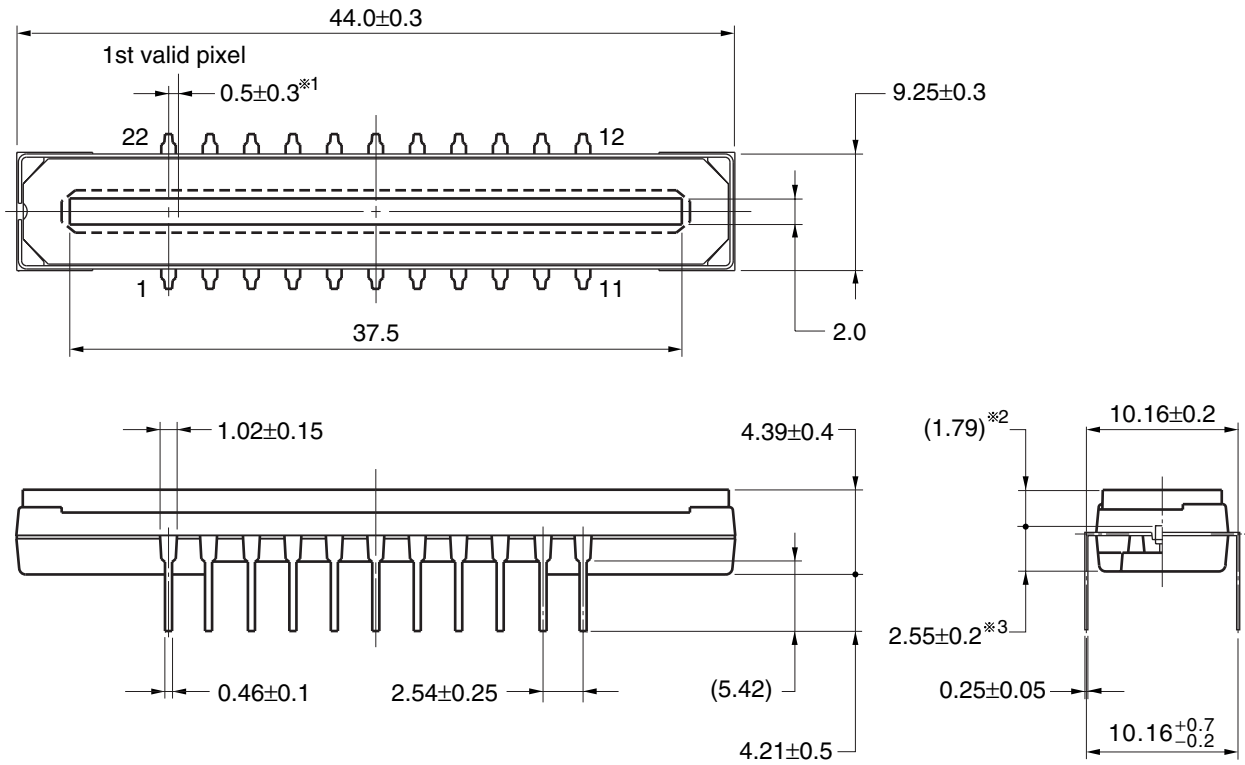
- Remarks 1.** The inverters shown in the above application circuit example are the 74HC04 (data rate < 2 MHz) or the 74AC04 (2 ≤ data rate < 10 MHz).
- 2.** Inverters B1 to B3 in the above application circuit example are shown in the figure below.



PACKAGE DRAWING

μPD8872CY
 CCD LINEAR IMAGE SENSOR 22-PIN PLASTIC DIP (10.16 mm (400))

(Unit : mm)



| Name | Dimensions | Refractive index |
|-------------|---------------|------------------|
| Plastic cap | 42.9×8.35×0.7 | 1.5 |

- ※1 1st valid pixel ↔ The center of the pin1
- ※2 The surface of the CCD chip ↔ The top of the cap
- ※3 The bottom of the package ↔ The surface of the CCD chip

22C-1CCD-PKG14-2

RECOMMENDED SOLDERING CONDITIONS

When soldering this product, it is highly recommended to observe the conditions as shown below.

If other soldering processes are used, or if the soldering is performed under different conditions, please make sure to consult with our sales offices.

Type of Through-hole Device

μPD8872CY-A : CCD linear image sensor 22-pin plastic DIP (10.16 mm (400))

| Process | Conditions |
|------------------------|--|
| Partial heating method | Pin temperature : 300 °C or below, Heat time : 3 seconds or less (per pin) |

- Cautions**
1. **During assembly care should be taken to prevent solder or flux from contacting the plastic cap. The optical characteristics could be degraded by such contact.**
 2. **Soldering by the solder flow method may have deleterious effects on prevention of plastic cap soiling and heat resistance. So the method cannot be guaranteed.**

NOTES ON HANDLING THE PACKAGES

① DUST AND DIRT PROTECTING

The optical characteristics of the CCD will be degraded if the cap is scratched during cleaning. Don't either touch plastic cap surface by hand or have any object come in contact with plastic cap surface. Should dirt stick to a plastic cap surface, blow it off with an air blower. For dirt stuck through electricity ionized air is recommended. And if the plastic cap surface is grease stained, clean with our recommended solvents.

○ CLEANING THE PLASTIC CAP

Care should be taken when cleaning the surface to prevent scratches.
 We recommend cleaning the cap with a soft cloth moistened with one of the recommended solvents below. Excessive pressure should not be applied to the cap during cleaning. If the cap requires multiple cleanings it is recommended that a clean surface or cloth be used.

○ RECOMMENDED SOLVENTS

The following are the recommended solvents for cleaning the CCD plastic cap.
 Use of solvents other than these could result in optical or physical degradation in the plastic cap. Please consult your sales office when considering an alternative solvent.

| Solvents | Symbol |
|----------------------|--------|
| Ethyl Alcohol | EtOH |
| Methyl Alcohol | MeOH |
| Isopropyl Alcohol | IPA |
| N-methyl Pyrrolidone | NMP |

② MOUNTING OF THE PACKAGE

The application of an excessive load to the package may cause the package to warp or break, or cause chips to come off internally. Particular care should be taken when mounting the package on the circuit board. Don't have any object come in contact with plastic cap. You should not reform the lead frame. We recommended to use a IC-inserter when you assemble to PCB.

Also, be care that the any of the following can cause the package to crack or dust to be generated.

1. Applying heat to the external leads for an extended period of time with soldering iron.
2. Applying repetitive bending stress to the external leads.
3. Rapid cooling or heating

③ OPERATE AND STORAGE ENVIRONMENTS

Operate in clean environments. CCD image sensors are precise optical equipment that should not be subject to mechanical shocks. Exposure to high temperatures or humidity will affect the characteristics. So avoid storage or usage in such conditions.

Keep in a case to protect from dust and dirt. Dew condensation may occur on CCD image sensors when the devices are transported from a low-temperature environment to a high-temperature environment. Avoid such rapid temperature changes.

For more details, refer to our document "Review of Quality and Reliability Handbook" (C12769E)

④ ELECTROSTATIC BREAKDOWN

CCD image sensor is protected against static electricity, but destruction due to static electricity is sometimes detected. Before handling be sure to take the following protective measures.

1. Ground the tools such as soldering iron, radio cutting pliers of or pincer.
2. Install a conductive mat or on the floor or working table to prevent the generation of static electricity.
3. Either handle bare handed or use non-chargeable gloves, clothes or material.
4. Ionized air is recommended for discharge when handling CCD image sensor.
5. For the shipment of mounted substrates, use box treated for prevention of static charges.
6. Anyone who is handling CCD image sensors, mounting them on PCBs or testing or inspecting PCBs on which CCD image sensors have been mounted must wear anti-static bands such as wrist straps and ankle straps which are grounded via a series resistance connection of about 1 MΩ.

[MEMO]

NOTES FOR CMOS DEVICES

① VOLTAGE APPLICATION WAVEFORM AT INPUT PIN

Waveform distortion due to input noise or a reflected wave may cause malfunction. If the input of the CMOS device stays in the area between V_{IL} (MAX) and V_{IH} (MIN) due to noise, etc., the device may malfunction. Take care to prevent chattering noise from entering the device when the input level is fixed, and also in the transition period when the input level passes through the area between V_{IL} (MAX) and V_{IH} (MIN).

② HANDLING OF UNUSED INPUT PINS

Unconnected CMOS device inputs can be cause of malfunction. If an input pin is unconnected, it is possible that an internal input level may be generated due to noise, etc., causing malfunction. CMOS devices behave differently than Bipolar or NMOS devices. Input levels of CMOS devices must be fixed high or low by using pull-up or pull-down circuitry. Each unused pin should be connected to V_{DD} or GND via a resistor if there is a possibility that it will be an output pin. All handling related to unused pins must be judged separately for each device and according to related specifications governing the device.

③ PRECAUTION AGAINST ESD

A strong electric field, when exposed to a MOS device, can cause destruction of the gate oxide and ultimately degrade the device operation. Steps must be taken to stop generation of static electricity as much as possible, and quickly dissipate it when it has occurred. Environmental control must be adequate. When it is dry, a humidifier should be used. It is recommended to avoid using insulators that easily build up static electricity. Semiconductor devices must be stored and transported in an anti-static container, static shielding bag or conductive material. All test and measurement tools including work benches and floors should be grounded. The operator should be grounded using a wrist strap. Semiconductor devices must not be touched with bare hands. Similar precautions need to be taken for PW boards with mounted semiconductor devices.

④ STATUS BEFORE INITIALIZATION

Power-on does not necessarily define the initial status of a MOS device. Immediately after the power source is turned ON, devices with reset functions have not yet been initialized. Hence, power-on does not guarantee output pin levels, I/O settings or contents of registers. A device is not initialized until the reset signal is received. A reset operation must be executed immediately after power-on for devices with reset functions.

⑤ POWER ON/OFF SEQUENCE

In the case of a device that uses different power supplies for the internal operation and external interface, as a rule, switch on the external power supply after switching on the internal power supply. When switching the power supply off, as a rule, switch off the external power supply and then the internal power supply. Use of the reverse power on/off sequences may result in the application of an overvoltage to the internal elements of the device, causing malfunction and degradation of internal elements due to the passage of an abnormal current.

The correct power on/off sequence must be judged separately for each device and according to related specifications governing the device.

⑥ INPUT OF SIGNAL DURING POWER OFF STATE

Do not input signals or an I/O pull-up power supply while the device is not powered. The current injection that results from input of such a signal or I/O pull-up power supply may cause malfunction and the abnormal current that passes in the device at this time may cause degradation of internal elements.

Input of signals during the power off state must be judged separately for each device and according to related specifications governing the device.

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